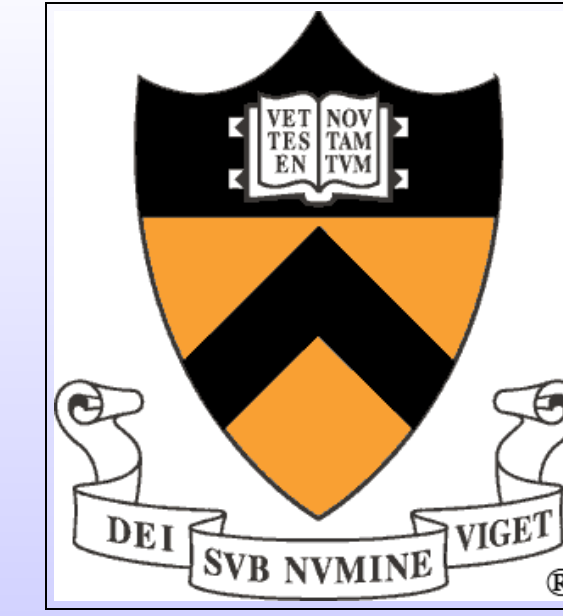




MEMORYLESS CHANNELS: THE BENEFITS OF FEEDBACK IN THE NON-ASYMPTOTIC REGIME

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Minimum energy per bit with and without feedback

Non-asymptotic problem statement

$$Z \sim \mathcal{N}(0, \frac{1}{2}N_0)$$

$$\downarrow$$

$$X \longrightarrow \oplus \longrightarrow Y$$

- (E, M, ϵ) code with feedback is given by a decoder $g: \mathbb{R}^\infty \rightarrow \{1, \dots, M\}$ and a sequence of encoder functions $f_j: \{1, \dots, M\} \times \mathbb{R}^{j-1} \rightarrow \mathbb{R}$, used to generate channel inputs:

$$X_j = f_j(W, Y^{j-1}),$$

and satisfying

$$\mathbb{E} \left[\sum_{j=1}^{\infty} X_j^2 \middle| W \right] \leq E, \quad \mathbb{P}[g(Y^\infty) \neq W] \leq \epsilon.$$

- (E, M, ϵ) code (without feedback) is required to satisfy $f_j(W, Y^{j-1}) = f_j(W)$.
- Energy-information tradeoff:

$$M^*(E, \epsilon) = \max\{M : \exists(M, E, \epsilon) \text{ code}\},$$

$$M_f^*(E, \epsilon) = \max\{M : \exists(M, E, \epsilon) \text{ code with feedback}\}$$

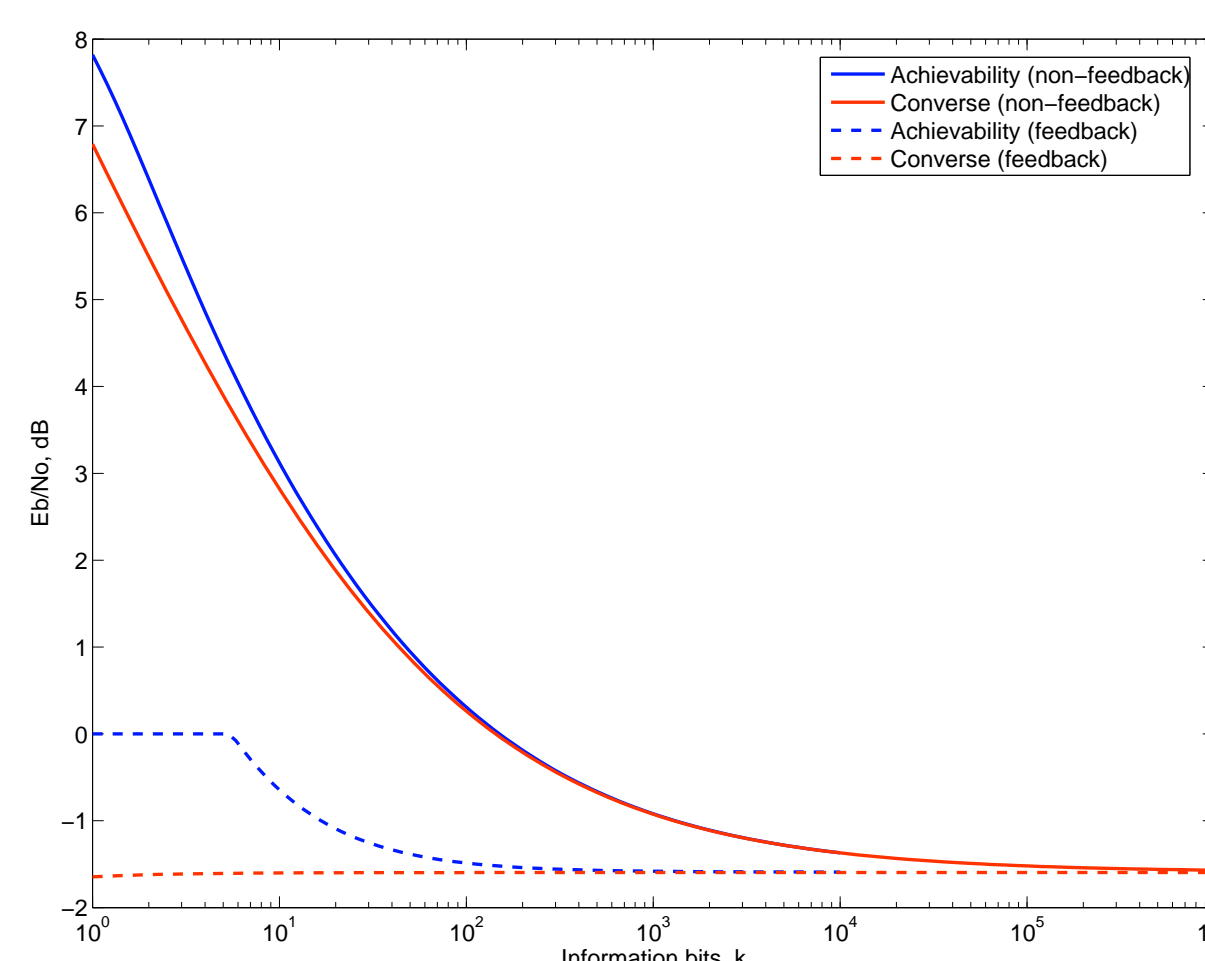
- Asymptotically we have [CS49]:

$$\lim_{E \rightarrow \infty} \frac{1}{E} \log M^*(E, \epsilon) = \lim_{E \rightarrow \infty} \frac{1}{E} \log M_f^*(E, \epsilon) = \frac{\log e}{N_0}, \quad 0 < \epsilon < 1$$

I.e. feedback does not improve minimum energy per bit.

- **Our work:** How do $M^*(E, \epsilon)$ and $M_f^*(E, \epsilon)$ compare for finite E ?

Results



Highlights:

- Horizontal axis: $k = \log_2 M$
- Without feedback: convergence to Shannon limit of -1.59 dB is slow, $O(\frac{1}{\sqrt{k}})$.
- With feedback: convergence is very fast.
- *Surprisingly*: decision feedback is enough!
- Plot: block error probability $\epsilon = 10^{-3}$; see [PPV10a].

- Without feedback:

$$\log M^*(E, \epsilon) = \frac{E}{N_0} \log e - \sqrt{\frac{2E}{N_0}} Q^{-1}(\epsilon) \log e + \frac{1}{2} \log \frac{E}{N_0} + O(1), \quad \epsilon > 0$$

$$\log M^*(E, 0) = 0$$

- With feedback:

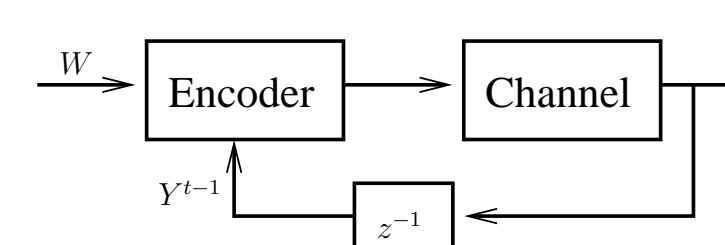
$$\log M_f^*(E, \epsilon) = \frac{E \log e}{N_0(1-\epsilon)} + O\left(\log \frac{E}{N_0}\right), \quad \epsilon > 0$$

$$\log M_f^*(E, 0) \geq \left\lfloor \frac{E}{N_0} \right\rfloor$$

- **Note:** feedback makes zero-error communication possible.

Variable-length coding with feedback

Non-asymptotic problem statement



- (ℓ, M, ϵ) variable-length feedback (VLF) code: sequences of encoder f_n and decoder functions g_n and a stopping time τ of filtration $\sigma\{Y^n\}$ such that:

$$\mathbb{P}[g_\tau(Y^\tau) \neq W] \leq \epsilon, \quad \mathbb{E}[\tau] \leq \ell,$$

where distribution P_{WY^n} is given by

$$P_{WY^n}(w, y^n) = \frac{1}{M} \prod_{j=1}^n P_{Y^j|X}(y_j | f_j(w, y^{j-1})).$$

- Non-asymptotic fundamental limit:

$$M^*(\ell, \epsilon) = \max\{M : \exists(\ell, M, \epsilon) \text{ -VLF code}\}.$$

- Burnashev [MB76] has shown:

$$\lim_{\ell \rightarrow \infty} \frac{1}{\ell} \log M^*(\ell, \exp\{-\ell E\}) = C \left(1 - \frac{E}{C_1}\right),$$

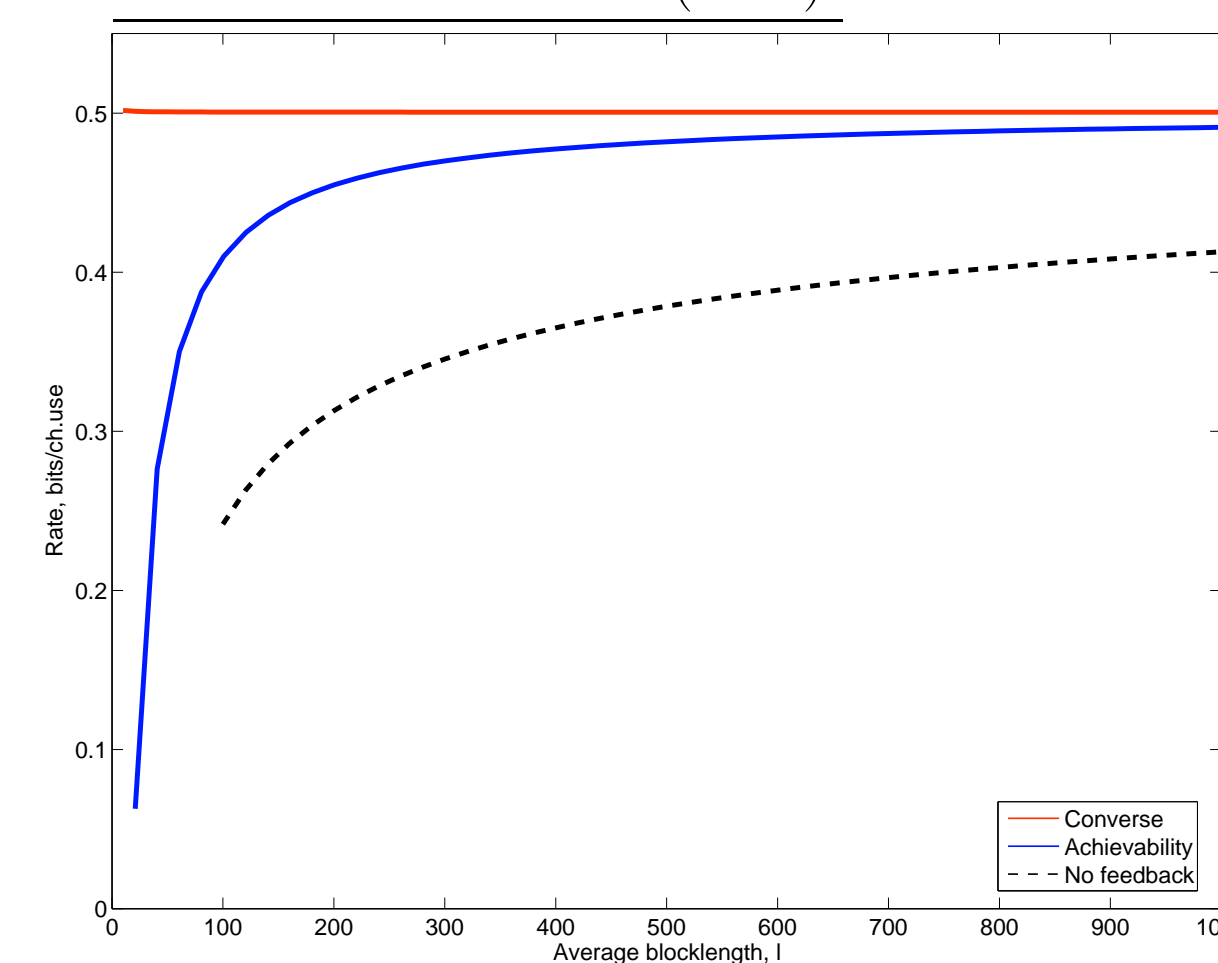
where C is the capacity and

$$C_1 = \max_{a_1, a_2 \in \mathcal{A}} D(P_{Y|X=a_1} \| P_{Y|X=a_2}).$$

- **Our work:** What is the behavior of $M^*(\ell, \epsilon)$ for a fixed ϵ ?

Results

VLF codes over BSC(0.11).



Highlights:

- Without feedback: convergence to capacity is slow, $O(\frac{1}{\sqrt{\ell}})$.
- VLF codes: *dispersion is zero*, i.e. convergence is much faster.
- Achievability (blue) bound: decision feedback only.
- Plot: probability of block error $\epsilon = 10^{-3}$.

- Main result [PPV10b]:

$$\log M_f^*(\ell, \epsilon) = \frac{\ell C}{1-\epsilon} + O(\log \ell), \quad 0 < \epsilon < 1$$

- Recall that without feedback [PPV10c]:

$$\log M^*(\ell, \epsilon) = \ell C - \sqrt{\ell V} Q^{-1}(\epsilon) + O(\log \ell),$$

where V is the channel dispersion.

Variable-length coding with termination

- **Problem:** Burnashev model assumes that control bits have the same reliability as information ones: in practice start/end is handled by upper layers.
- **Modified model:** “use-once” noiseless termination symbol.

$$010110T \rightarrow \text{[Channel]} \rightarrow 100110TTTTT \dots$$

- **VLFT code:** a VLF code that employs $T \iff \tau$ is a stopping time of filtration $\sigma\{W, Y^n\}, n = 1, \dots$

$$M_t^*(\ell, \epsilon) = \max\{M : \exists(\ell, M, \epsilon) \text{ VLFT code}\}.$$

- Examples of VLFT: ARQ, fountain codes.
- **Question:** Does $M_t^*(\ell, \epsilon)$ differ significantly from $M_f^*(\ell, \epsilon)$?

Results

- Without termination: for all channels with $C_1 < \infty$ (e.g. BSC) we have:

$$\text{VLF} : \log M_f^*(\ell, 0) = 0$$

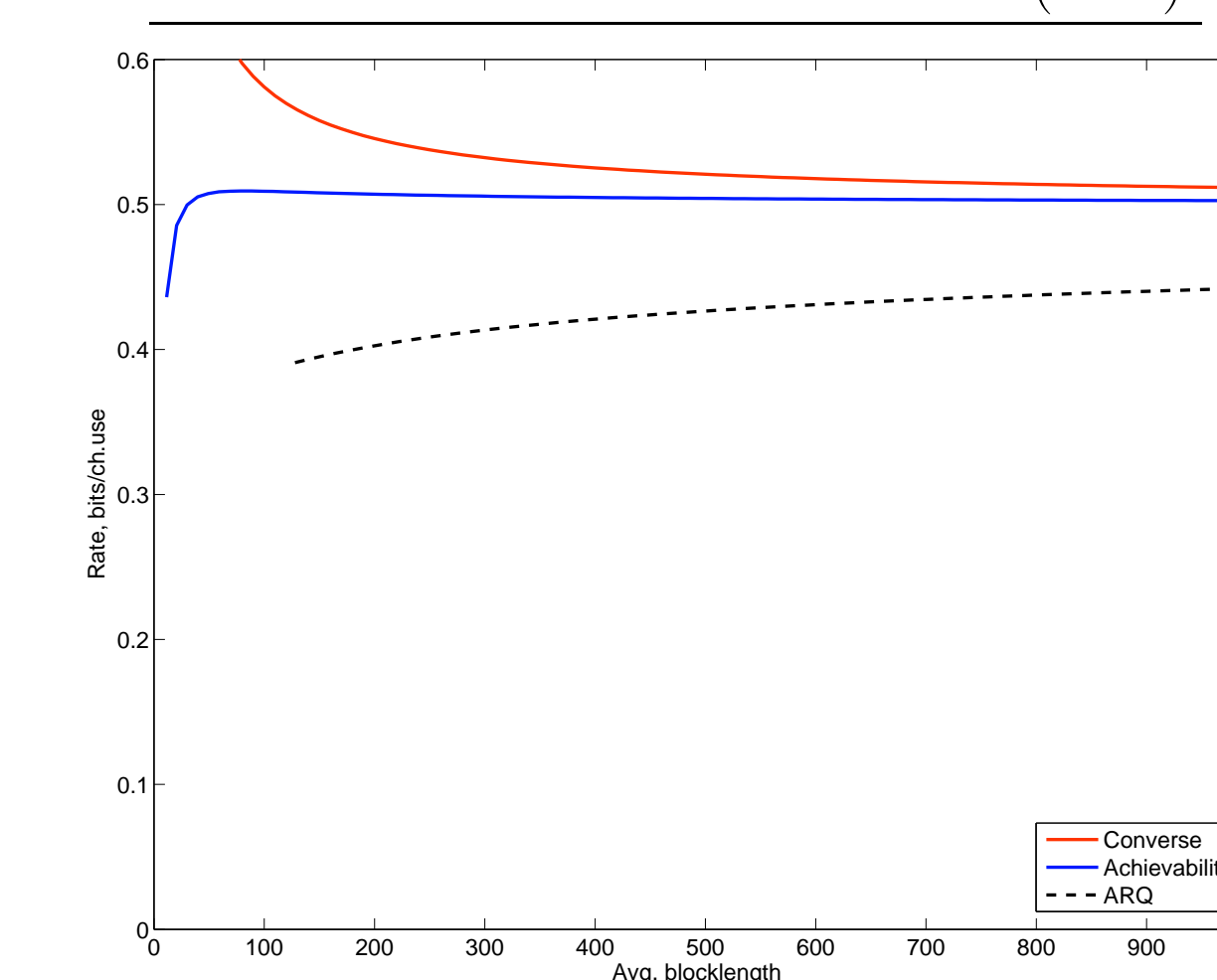
- With termination: ARQ paired with optimal fixed-length code achieves capacity [PPV10c]:

$$\text{ARQ} : \log M_t^*(\ell, 0) \geq \ell C - \text{const} \sqrt{\ell \log \ell}$$

- But we can do much better [PPV10b]:

$$\text{VLFT} : \log M_t^*(\ell, 0) \geq \ell C + O(1)$$

Zero-error VLFT codes over BSC(0.11).



Highlights:

- ARQ: very slow convergence even when the best possible block codes are used.
- New codes: capacity is achievable at *very short code-length* (and with zero error).

- In summary, the minimal blocklength to achieve $R = 0.9C$ for $BSC(0.11)$:

fixed-length : $\ell \approx 3100$	penalty term: $O(\sqrt{\ell})$
VLF : $\ell \lesssim 200$	penalty term: $O(\log \ell)$
VLF + termination : $\ell \lesssim 20$	penalty term: $O(1)$

References

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